



### Applications

- Low voltage, high density systems with Intermediate Bus Architectures (IBA)
- Point-of-load regulators for high performance DSP, FPGA, ASIC, and microprocessor applications
- Industrial computing, servers, and storage
- Broadband, networking, optical, and wireless communications systems
- Active memory bus terminators

### Benefits

- Integrates digital power conversion with intelligent power management
- Eliminates the need for external power management components and communication bus
- Completely programmable via pin strapping and one external resistor
- One part that covers all applications
- Reduces board space, system cost and complexity, and time to market

### Description

Power-One's point-of-load converters are recommended for use with regulated bus converters in an Intermediate Bus Architecture (IBA). The ZY2140 is an intelligent, fully programmable step-down point-of-load DC-DC module integrating digital power conversion and power management. The ZY2140 completely eliminates the need for external components for sequencing, tracking, protection, monitoring, and reporting. Performance parameters of the ZY2140 are programmable by pin strapping and an external resistor and can be changed by the user at any time during product development and service without a need for a communication bus.

### Reference Documents

No-Bus<sup>™</sup> POL Converters. Application Note  
Z-One<sup>®</sup> POL Converters. Eutectic Solder Process Application Note  
Z-One<sup>®</sup> POL Converters. Lead-Free Process Application Note

### Features

- RoHS lead free and lead-solder-exempt products are available
- Wide input voltage range: 8V–14V
- High continuous output current: 40A
- Wide programmable output voltage range: 0.5V–3.65V
- Active digital current share
- Output voltage margining
- Overcurrent and overtemperature protections
- Overvoltage and undervoltage protections, and Power Good signal tracking the output voltage setpoint
- Tracking during turn-on and turn-off with guaranteed slew rates
- Sequenced and cascaded modes of operation
- Single-wire line for frequency synchronization between multiple POLs
- Programmable feedback loop compensation
- Differential output voltage sense
- Enable control
- Flexible fault management and propagation
- Start-up into the load pre-biased up to 100%
- Current sink capability
- Real time current measurements, monitoring, and reporting
- Industry standard size through-hole single-in-line package: 1.8"x0.55"
- Low height of 1.1"
- Wide operating temperature range: 0 to 70°C
- UL60950 recognized, CSA C22.2 No. 60950-00 certified, and TUV EN60950-1:2001 certified (pending)

## 1. Ordering Information

ZY	21	40	y	-	zz
<b>Product family:</b> Z-One Module	<b>Series:</b> No-Bus POL Converter	<b>Output Current:</b> 40A	<b>RoHS compliance:</b> <b>No suffix</b> - RoHS compliant with Pb solder exemption <sup>1</sup> <b>G</b> - RoHS compliant for all six substances	<b>Dash</b>	<b>Packaging Option<sup>2</sup>:</b> <b>R1</b> – 30 pcs Tray <b>Q1</b> – 1 pc sample for evaluation only

<sup>1</sup> The solder exemption refers to all the restricted materials except lead in solder. These materials are Cadmium (Cd), Hexavalent chromium (Cr6+), Mercury (Hg), Polybrominated biphenyls (PBB), Polybrominated diphenylethers (PBDE), and Lead (Pb) used anywhere except in solder.

<sup>2</sup> Packaging option is used only for ordering and not included in the part number printed on the POL converter label.

Example: **ZY2140G-R3**: A 30-piece tray of RoHS compliant POL converters. Each POL converter is labeled ZY2140G.

## 2. Absolute Maximum Ratings

Stresses in excess of the absolute maximum ratings may cause performance degradation, adversely affect long-term reliability, and cause permanent damage to the POL converter.

Parameter	Conditions/Description	Min	Max	Units
Operating Temperature	Controller Case Temperature	-40	105	°C
Input Voltage	250ms Transient		15	VDC

## 3. Environmental and Mechanical Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Ambient Temperature Range		0		70	°C
Storage Temperature (Ts)		-55		125	°C
Weight			33		grams
Operating Vibration (sinusoidal)	Frequency Range Magnitude Sweep Rate Repetitions in each axis (Min-Max-Min Sweep)	5 0.5 1 2		500	Hz G oct/min sweeps
Non-Operating Shock (half sine)	Acceleration Duration Number of shocks in each axis	50 11 10			G ms
MTBF	Calculated Per Telcordia Technologies SR-332	TBD			MHrs
Peak Reflow Temperature	ZY2140			220	°C
Peak Reflow Temperature	ZY2140G		245	260	°C
Lead Plating	ZY2140 and ZY2140G	100% Matte Tin			
Moisture Sensitivity Level	JEDEC J-STD-020C	3			

#### 4. Electrical Specifications

Specifications apply at the input voltage from 8V to 14V, output load from 0 to 40A, ambient temperature from 0°C to 70°C, output capacitance consisting of 110µF ceramic and 220µF tantalum, and the CCA=1 unless otherwise noted.

##### 4.1 Input Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Input voltage ( $V_{IN}$ )	Refer to Figure 1	8		14	VDC
Undervoltage Lockout Threshold	Ramping Up		7.5		VDC
	Ramping Down		6.6		VDC
Input Current	$V_{IN}=12V$		23		mADC
Maximum Input Current	$V_{IN}=8V$			19.7	ADC

##### 4.2 Output Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Output Current ( $I_{OUT}$ )	$V_{IN\ MIN}$ to $V_{IN\ MAX}$	-40 <sup>1</sup>		40	ADC
Output Voltage Range ( $V_{OUT}$ )	Programmable <sup>2</sup> with a resistor between TRIM and MARGIN pins Default (no resistor)	0.5		3.65	VDC
			0.5		VDC
Output Voltage Setpoint Accuracy <sup>3</sup>	$V_{IN}=12V$ , $I_{OUT}=0.5*I_{OUT\ MAX}$ , room temperature	±1.5% or 20mV whichever is greater			% $V_{OUT}$
Line Regulation <sup>3</sup>	$V_{IN\ MIN}$ to $V_{IN\ MAX}$		±0.5		% $V_{OUT}$
Load Regulation <sup>3</sup>	0 to $I_{OUT\ MAX}$		±0.5		% $V_{OUT}$
Dynamic Regulation Peak Deviation Settling Time	50% - 75% - 50% load step Slew rate 1A/µs, $C_{OUT}=380\mu F$ to 10% of peak deviation		280		mV
			100		µs
Output Voltage Peak-to-Peak Ripple and Noise BW=20MHz Full Load	$V_{IN}=12V$ , $V_{OUT}=0.75V$ $V_{IN}=12V$ , $V_{OUT}=2.5V$ $V_{IN}=12V$ , $V_{OUT}=3.3V$		15		mV
			25		mV
			35		mV
Efficiency $F_{SW}=500kHz$ Full Load Room temperature	$V_{IN}=12V$ , $V_{OUT}=0.5V$ $V_{IN}=12V$ , $V_{OUT}=0.75V$ $V_{IN}=12V$ , $V_{OUT}=1.0V$ $V_{IN}=12V$ , $V_{OUT}=1.2V$ $V_{IN}=12V$ , $V_{OUT}=1.8V$ $V_{IN}=12V$ , $V_{OUT}=2.5V$ $V_{IN}=12V$ , $V_{OUT}=3.3V$		75.3		%
			81.9		%
			85.3		%
			87		%
			90.2		%
			92.1		%
			93.5		%
Temperature Coefficient	$V_{IN}=12V$ , $I_{OUT}=0.5*I_{OUT\ MAX}$		35		ppm/°C
Switching Frequency	2 phases combined	450	500	550	kHz

<sup>1</sup> At the negative output current (bus terminator mode) efficiency of the ZY2140 degrades resulting in increased internal power dissipation. Maximum allowable negative current is limited to 40A.

<sup>2</sup> ZY2140 is a step-down converter, thus the output voltage is always lower than the input voltage as show in **Error! Reference source not found..**

<sup>3</sup> Digital PWM has an inherent quantization uncertainty of ±6.25mV that is not included in the specified static regulation parameters.

### 4.3 Protection Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
<b>Output Overcurrent Protection</b>					
Type		Non-Latching, 130ms period			
Threshold			140		%I <sub>OUT</sub>
Threshold Accuracy		-25		25	%I <sub>OCP.SET</sub>
<b>Output Overvoltage Protection</b>					
Type		Latching			
Threshold	Follows the output voltage setpoint		130 <sup>1</sup>		%V <sub>O.SET</sub>
Threshold Accuracy	Measured at V <sub>O.SET</sub> =2.5V	-2		2	%V <sub>OVP.SET</sub>
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
<b>Output Undervoltage Protection</b>					
Type		Non-Latching, 130ms period			
Threshold	Follows the output voltage setpoint		75		%V <sub>O.SET</sub>
Threshold Accuracy	Measured at V <sub>O.SET</sub> =2.5V	-2		2	%V <sub>UVP.SET</sub>
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
<b>Overtemperature Protection</b>					
Type		Non-Latching, 130ms period			
Turn Off Threshold	Temperature is increasing		120		°C
Turn On Threshold	Temperature is decreasing after module was shut down by OTP		110		°C
Threshold Accuracy		-5		5	°C
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
<b>Power Good Signal (PGOOD pin)</b>					
Logic	V <sub>OUT</sub> is inside the PG window and stable V <sub>OUT</sub> is outside of the PG window or ramping up/down		High Low		N/A
Lower Threshold	Follows the output voltage setpoint		90		%V <sub>O.SET</sub>
Upper Threshold	Follows the output voltage setpoint		110		%V <sub>O.SET</sub>
Delay	From instant when threshold is exceeded until status of PG pin changes		6		µs
Threshold Accuracy	Measured at V <sub>O.SET</sub> =2.5V	-2		2	%V <sub>O.SET</sub>

<sup>1</sup> Minimum OVP threshold is 1.0V

#### 4.4 Feature Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
<b>Current Share (CS pin)</b>					
Type		Active, Single Line			
Current Share Accuracy	$I_{OUT\ MIN} \geq 20\% * I_{OUT\ NOM}$			±20	% $I_{OUT}$
<b>Tracking</b>					
Rising Slew Rate	Proportional to SYNC frequency		0.1		V/ms
Falling Slew Rate	Proportional to SYNC frequency		-0.5		V/ms
<b>Enable (EN pin)</b>					
EN Pin Polarity		Positive (enables the output when EN pin is open or pulled high)			
EN High Threshold		2.3			VDC
EN Low Threshold				1.0	VDC
Open Circuit Voltage			3.3		VDC
Turn-On Delay	From EN pin changing state to $V_{OUT}$ starting to ramp up		0		ms
Turn-Off Delay	From EN pin changing state to $V_{OUT}$ reaching 0V		11		ms
<b>Feedback Loop Compensation (CCA pin)</b>					
CCA pin is open	Recommended $C_{OUT}$ /ESR range, combination of ceramic + tantalum	50/5 + 220/40	100/5 + 470/40	400/5 + 2000/20	$\mu$ F/m $\Omega$ $\mu$ F/m $\Omega$
CCA pin is connected to GND	Recommended $C_{OUT}$ /ESR range, ceramic	100/5	220/5	400/5	$\mu$ F/m $\Omega$
<b>Output Current Monitoring (CS pin)</b>					
Output Current Monitoring Accuracy	$30\% * I_{OUT\ NOM} < I_{OUT} < I_{OUT\ NOM}$ $V_{IN} = 12V$	-20		+20	% $I_{OUT}$
Conversion Ratio	Duty Cycle of the negative pulse corresponding to 100% of nominal current		74		%
<b>Remote Voltage Sense (-VS and +VS pins)</b>					
Type		Differential			
Voltage Drop Compensation	Between +VS and $V_{OUT}$			300	mV
Voltage Drop Compensation	Between -VS and PGND			100	mV

#### 4.5 Signal Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
VDD	Internal supply voltage	3.15	3.3	3.45	V
<b>SYNC Line</b>					
ViL_s	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_s	HIGH level input voltage	0.75 x VDD		VDD + 0.5	V
Vhyst_s	Hysteresis of input Schmitt trigger	0.25 x VDD		0.45 x VDD	V
IoL_s	LOW level sink current V(SYNC)=0.5V	14		60	mA
Ipu_s	Pull-up current source V(SYNC)=0V	300		1000	μA
Tr_s	Maximum allowed rise time 10/90%VDD			300	ns
Cnode_s	Added node capacitance		5	10	pF
Freq_s	Clock frequency of external SYNC line	475		525	kHz
Tsynq	Sync pulse duration	22		28	% of clock cycle
T0	Data=0 pulse duration	72		78	% of clock cycle
<b>Inputs: CCA, EN, IM</b>					
Iup_x	Pull-up current source V(X)=0	25		110	μA
ViL_x	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_x	HIGH level input voltage	0.7 x VDD		VDD+0.5	V
Vhyst_x	Hysteresis of input Schmitt trigger	0.1 x VDD		0.3 x VDD	V
RdnL_x	External pull down resistance pin forced low			10	kΩ
<b>Power Good and OK Inputs/Outputs</b>					
Iup_PG	Pull-up current source V(PG)=0	25		110	μA
Iup_OK	Pull-up current source V(OK)=0	175		725	μA
ViL_x	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_x	HIGH level input voltage	0.7 x VDD		VDD+0.5	V
Vhyst_x	Hysteresis of input Schmitt trigger	0.1 x VDD		0.3 x VDD	V
IoL_x	LOW level sink current at 0.5V	4		20	mA
<b>Current Share/Sense Bus</b>					
Iup_CS	Pull-up current source at V(CS)=0V	0.84		3.10	mA
ViL_CS	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_CS	HIGH level input voltage	0.75 x VDD		VDD+0.5	V
Vhyst_CS	Hysteresis of input Schmitt trigger	0.25 x VDD		0.45 x VDD	V
IoL_CS	LOW level sink current V(CS)=0.5V	14		60	mA
Tr_CS	Maximum allowed rise time 10/90% VDD			100	ns

**5. Typical Performance Characteristics**

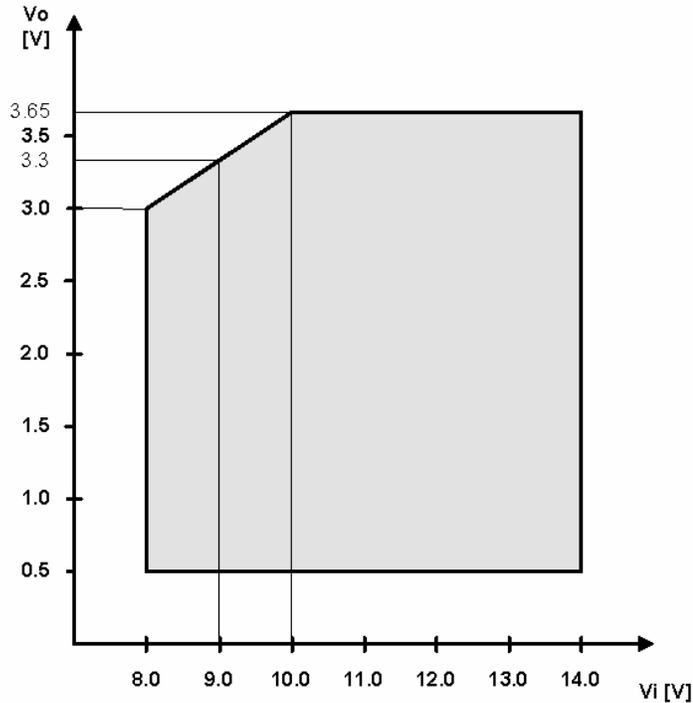


Figure 1. Output Voltage as a Function of Input Voltage

**5.1 Efficiency Curves**

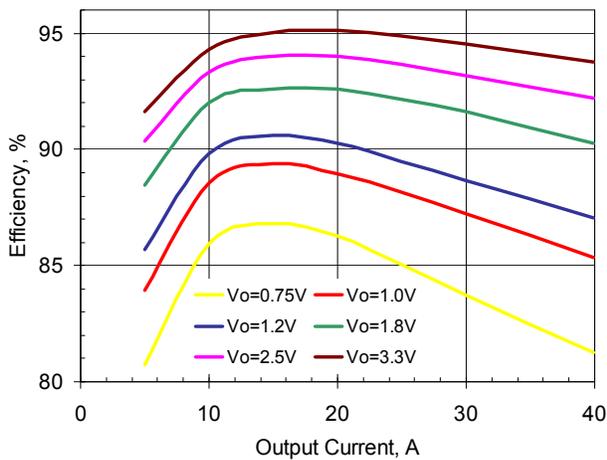


Figure 2. Efficiency vs. Load. Vin=9.6V

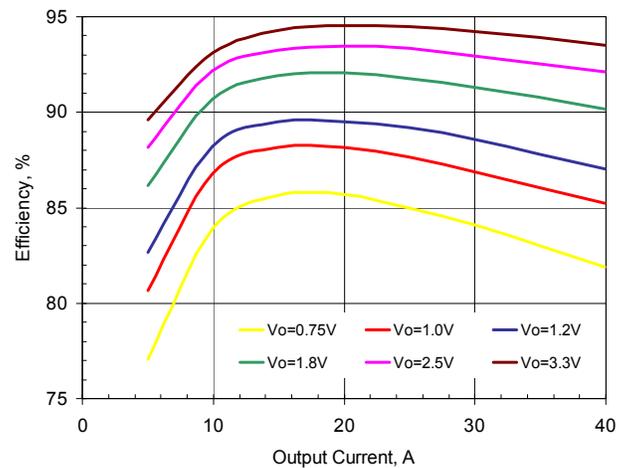


Figure 3. Efficiency vs. Load. Vin=12V

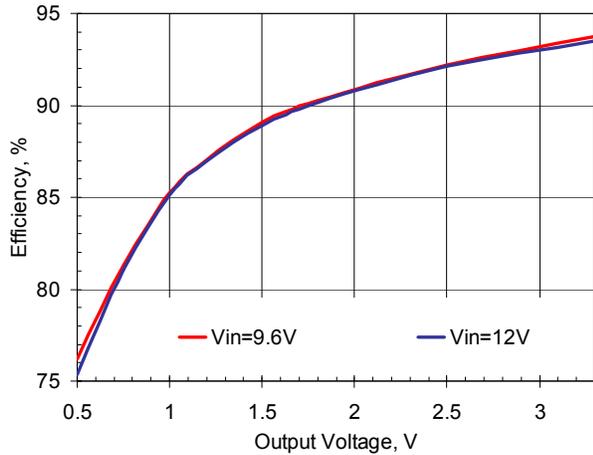


Figure 4. Efficiency vs. Output Voltage, Iout=40A

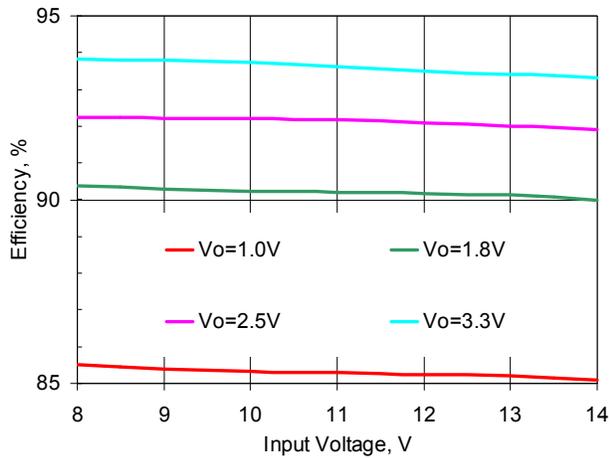


Figure 5. Efficiency vs. Input Voltage, Iout=40A

### 5.2 Turn-On Characteristics

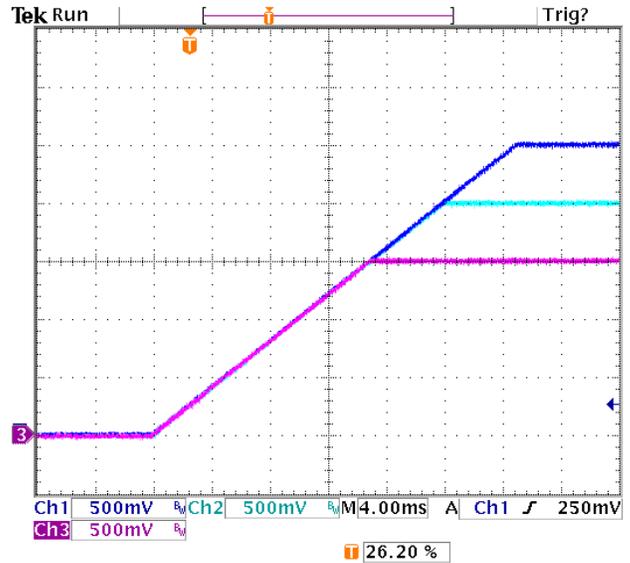


Figure 6. Tracking Turn-On.  
Vin=12V, Ch1 – V1, Ch2 – V2, Ch3 – V3

### 5.3 Turn-Off Characteristics

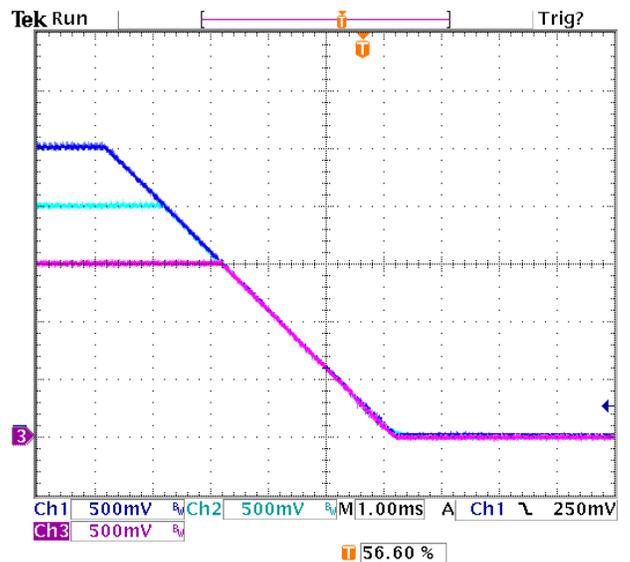


Figure 7. Tracking Turn-Off  
Vin=12V, Ch1 – V1, Ch2 – V2, Ch3 – V3

### 5.4 Transient Response

The pictures below show the deviation of the output voltage in response to the 50%-75%-50% step load at 1.0A/μs. In all tests the POL converter had a total of 380μF ceramic and tantalum capacitors connected across the output pins. The speed of the transient response was varied by selecting different CCA settings.

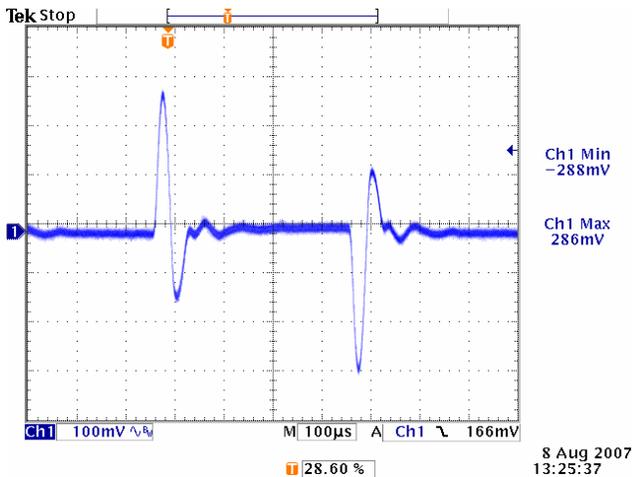


Figure 8. Vin=12V, Vout=3.3V. CCA=1

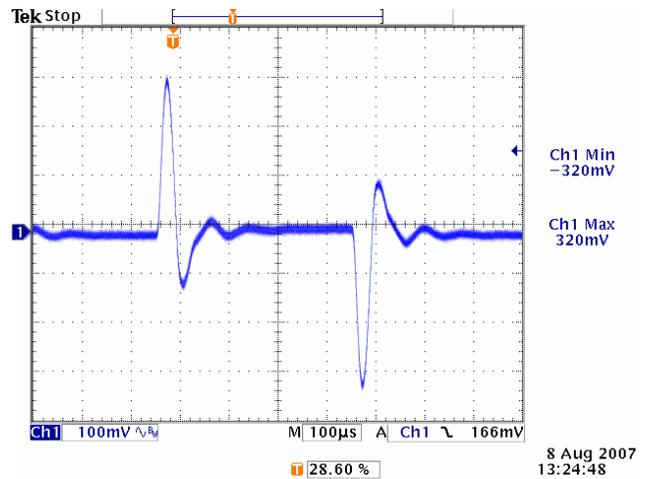


Figure 9. Vin=12V, Vout=3.3V. CCA=0

### 5.5 Thermal Derating Curve

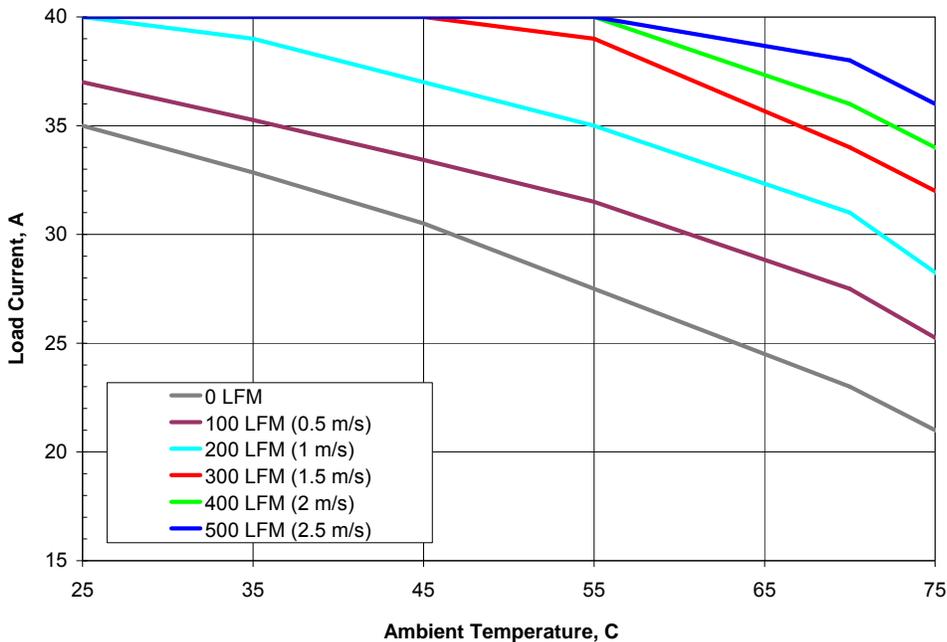


Figure 10. Thermal Derating Curves. Vin=12V, Vout=3.3V

6. Typical Application

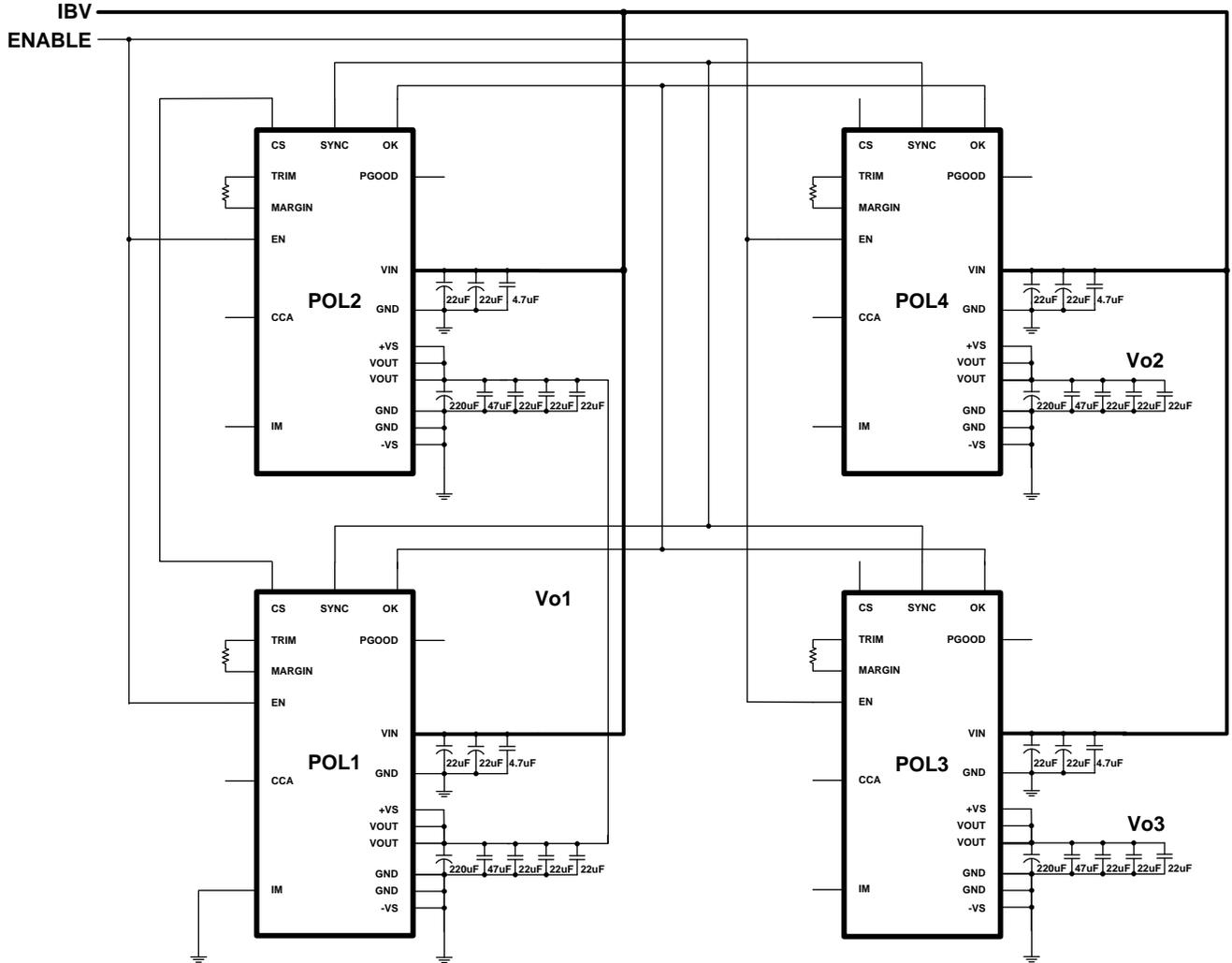


Figure 11. Complete Schematic of Application with Three Independent Outputs. Intermediate Bus Voltage is from 8V to 14V.

In this application four POL converters are configured to deliver three independent output voltages. POL1 and POL2 are connected in parallel for increased output current. Output voltages are programmed with the resistors connected between TRIM and MARGIN pins of individual converters.

POL1 is configured as a master (IM pin is grounded) and all other POL converters are synchronized to the switching frequency of POL1.

All converters are controlled by the common ENABLE signal. Turn-on and turn-off processes of the system are illustrated by pictures in Figure 6 and Figure 7.

## 7. Pin Assignments and Description

Pin Name	Pin Number	Pin Type	Buffer Type	Pin Description	Notes
OK	6	I/O	PU	Fault Status	Connect to OK pin of other Z- POLs. Leave open, if not used
SYNC	4	I/O	PU	Frequency Synchronization Line	Connect to SYNC pin of other Z-POLs or to an external clock generator
PGOOD	11	I/O	PU	Power Good	
CS	3	I/O	PU	Current Share/Sense	Connect to CS pin of other Z-POLs connected in parallel
IM	9	I	PU	Master Mode	Tie to GND to make the POL the clock master or leave open to synchronize to external clock
CCA	10	I	PU	Compensation Coefficient Address	Tie to GND for 0 or leave open for 1
MARGIN	8	A		Output Voltage Margining	To program the output voltage, connect a resistor between MARGIN and TRIM
EN	5	I	PU	Enable	POL is ON when the pin is high or floating. POL is OFF when the pin is low or connected to GND
TRIM	7	A		Output Voltage Trim	To program the output voltage, connect a resistor between MARGIN and TRIM
-VS	18	I	A	Negative Voltage Sense	Connect to the negative point close to the load
+VS	15	I	A	Positive Voltage Sense	Connect to the positive point close to the load
VOUT	12, 13	P		Output Voltage	
GND	2, 16, 17	P		Power Ground	
VIN	1	P		Input Voltage	

Legend: I=input, O=output, I/O=input/output, P=power, A=analog, PU=internal pull-up

## **8. Pin and Feature Description**

### **8.1 OK, Fault Status**

The open drain input/output with the internal pull-up resistor. The POL converter pulls its OK pin low, if a fault occurs. Pulling low the OK input by an external circuitry turns off the POL converter.

### **8.2 SYNC, Frequency Synchronization Line**

The bidirectional input/output with the internal pull-up resistor. If the POL converter is configured as a master, the SYNC line propagates clock to other POL converters. If the POL converter is configured as a slave, the internal clock recovery circuit synchronizes the POL converter to the clock of the SYNC line.

### **8.3 IM, Interleave Mode**

The input with the internal pull-up resistor. When the pin is left floating, the switching frequency is determined by an external clock applied to the SYNC pin. Pulling the IM pin low configures a POL converter as a master. The master determines the clock on the SYNC line.

### **8.4 PG, Power Good**

The open drain input/output with the internal pull-up resistor. The pin is pulled low by the POL converter, if the output voltage is outside of the window defined by the Power Good High and Low thresholds.

### **8.5 CCA, Compensation Coefficient Address**

Inputs with internal pull-ups to select one of 2 sets of digital filter coefficients optimized for different characteristics of output capacitance.

### **8.6 CS, Current Share/Sense Bus**

The open drain digital input/output with the internal pull-up resistor. The duty cycle of the digital signal is proportional to the output current of the POL converter. External capacitive loading of the pin shall be avoided.

### **8.7 MARGIN, Output Voltage Margining**

The output of the 2V internal voltage reference that is used to program the output voltage of the POL converter.

### **8.8 TRIM, Output Voltage Trim**

The input of the TRIM comparator for the output voltage programming.

The output voltage is programmed by a single resistor connected between MARGIN and TRIM pins.

### **8.9 EN, Enable**

The input with the internal pull-up resistor. The POL converter is turned off, when the pin is pulled low

### **8.10 -VS and +VS**

The differential voltage input of the POL converter feedback loop.

## 9. Application Information

### 9.1 Output Voltage Programming

Resistance of the trim resistor is determined from the equation below:

$$R_{TRIM} = \frac{20 \times (5.5 - V_{OUT})}{V_{OUT}}, \text{ k}\Omega$$

where  $V_{OUT}$  is the desired output voltage in Volts.

If the  $R_{TRIM}$  is open or the TRIM pin is shorted to PGND, the  $V_{OUT}=0.5V$ .

### 9.2 Output Voltage Margining

Margining can be implemented by changing the resistance between the REF and TRIM pins.

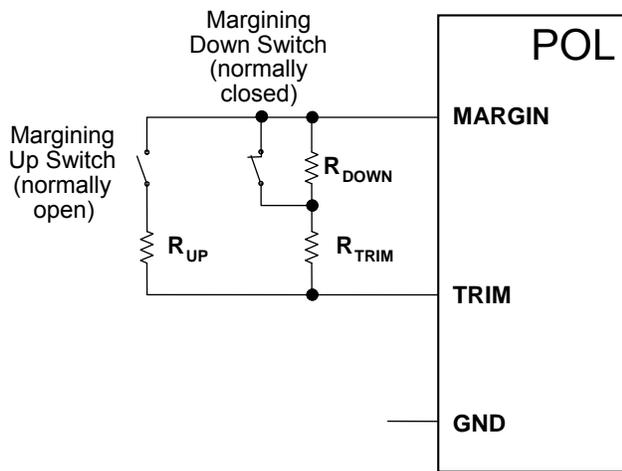


Figure 12. Margining Configuration

In the schematic shown in Figure 12, the nominal output voltage is set with the trim resistor  $R_{TRIM}$  calculated from the equation in the paragraph 9.1. Resistors  $R_{UP}$  and  $R_{DOWN}$  are added to margin the output voltage up and down respectively and determined from the equations below.

$$R_{UP} = \frac{20 \times R_{TRIM}}{20 + R_{TRIM}} \times \left( \frac{5 \times R_{TRIM} - \Delta V\%}{\Delta V\%} \right), \text{ k}\Omega$$

$$R_{DOWN} = (20 + R_{TRIM}) \times \left( \frac{\Delta V\%}{100 - \Delta V\%} \right), \text{ k}\Omega$$

where  $R_{TRIM}$  is the value of the trim resistor in  $k\Omega$  and  $\Delta V\%$  is the absolute value of desired margining expressed in percents of the nominal output voltage.

During normal operation the resistors are removed from the circuit by the switches. The “Margining Down” switch is normally closed shorting the resistor  $R_{DOWN}$  while the “Margining Up” switch is normally open disconnecting the resistor  $R_{UP}$ .

An alternative configuration of the margining circuit is shown in Figure 13. In the configuration both switches are normally open that may be advantageous in some implementations.

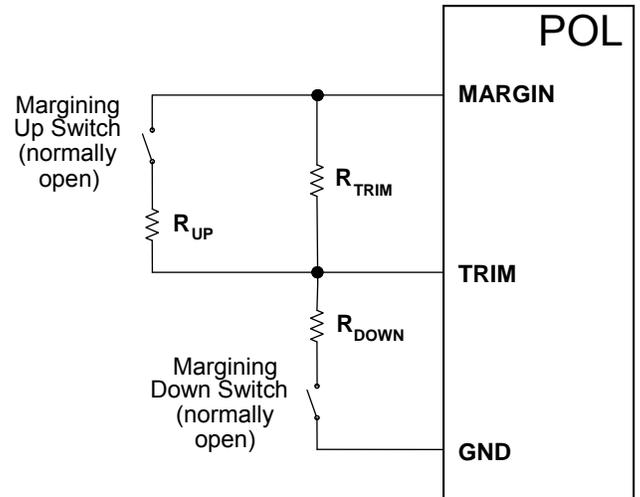


Figure 13. Alternative Margining Configuration

$R_{UP}$  and  $R_{DOWN}$  for this configuration are determined from the following equations:

$$R_{UP} = \frac{20 \times R_{TRIM}}{20 + R_{TRIM}} \times \left( \frac{5 \times R_{TRIM} - \Delta V\%}{\Delta V\%} \right), \text{ k}\Omega$$

$$R_{DOWN} = \frac{20 \times R_{TRIM}}{20 + R_{TRIM}} \times \left( \frac{100 - \Delta V\%}{\Delta V\%} \right), \text{ k}\Omega$$

**Caution:** Noise injected into the TRIM node may affect accuracy of the output voltage and stability of the POL converter. Always minimize the PCB trace length from the TRIM pin to external components to avoid noise pickup.

Refer to *No-Bus™ POL Converters. Application Note* on [www.power-one.com](http://www.power-one.com) for more application information on this and other product features.

